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GO DIGITAL



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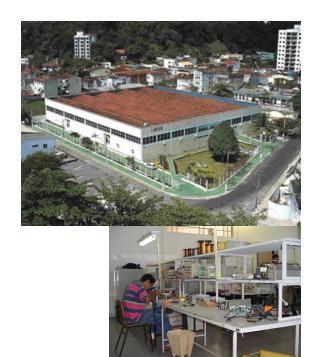


1. Introduction



Linear Corporate Overview

- USA factory created in 2005
- Local assembly
- Full test facility
- Marketing, Sales & Technical support
- Spare Parts inventory
- 30 Minutes from Chicago ORD Airport
- Most transmitters shipped in less than 10 days
- Linear in Brazil was established in 1978
- 28,000 Transmitter and Translator installations in 40 countries
- 350 Employees
- Largest supplier of Transmitters in South America and in 2006, 2007 and 2008 probably the world!
- Predominate supplier of digital TV equipment to the new South American digital market







"Flagship" products







DIGITAL EXCITER / TRANSMITTER

- VHF and UHF
- 20 120 Watts
- Transmitter or Exciter
- Air Cooled
- Linear and non-Linear correction
- Ideal for flash-cutting 1kW analog transmitters
- All metering on front panel
- **Transport Stream measurement**
- ASI or SMPTE 310M input
- Variable power control from front panel

DIGITAL EXCITER

- Frequency Agile
- 200 mW output
- Automatic linear and non-linear correction
- Ideal for flash-cuts works with ANY OEM transmitter
- Includes measurement software option
- 14 different ATSC measurements
- ASI or SMPTE 310 input
- One-button correction



- **UHF TV Transmitter**
- 15 10,000 Watts
- Transmitters and **Translators**
- Gap Fillers
- Air Cooled
- Analog NTSC and Digital **ATSC**
- Linear and non-linear correction
- Unparallel performance
- 2 years warranty



Summary Of ATSC standard

- ATSC baseband signal is an MPEG-2 transport stream (TS)
 - Video signal coded as MPEG-2 SD or HD
 - Audio signal coded to the Dolby Labs AC-3 format / 5.1 Audio
- 8VSB modulation scheme is a single-carrier technique employing eight-level trellis coding
- I (Real) Q (imaginary) modulation
- 8 VSB uses I axis with 8 level modulation constellation points equidistantly distributed
- Bandwidth reduced by lower sideband suppression (removing some of the Q component)
- As in analog TV VSB filtering resulting; upper and vestigial sideband is employed
- Receiver contains Nyquist filtering to improve sideband filtering effectiveness
- Inherent 8 VSB advantage is power efficiency: ATSC Digital apparently uses only 25% power of NTSC analog signal to cover the same area



Main Causes Of Bit Errors in Transport Stream

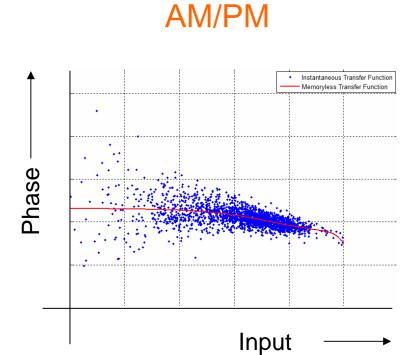
- Echoes* External to Transmission system
- Interferers* External to transmission system
- Amplitude distortion typically non-linear amplifiers
- IQ errors amplitude and phase errors in the modulator
- Group Delay errors typically linear phase distortion
- <u>Phase jitter</u> Poor oscillator stability, encoder phase errors or translator receiver phase errors
- <u>Power spectrum</u> "Shoulder" levels too high Insufficient shoulder attenuation through non-linear correction and mask filter attenuation

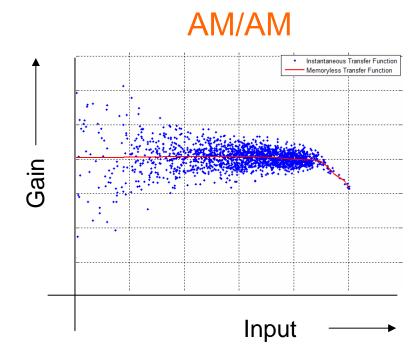


^{*} Not covered in this presentation



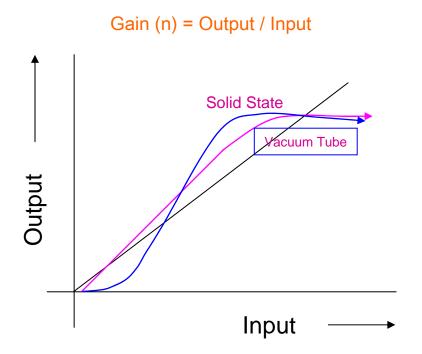
Amplitude And Phase Distortion

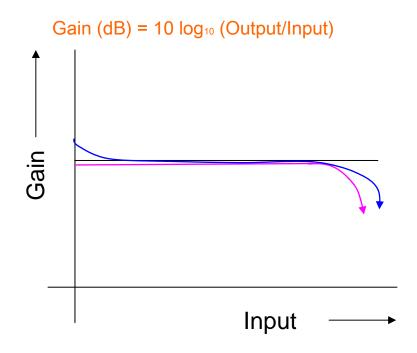






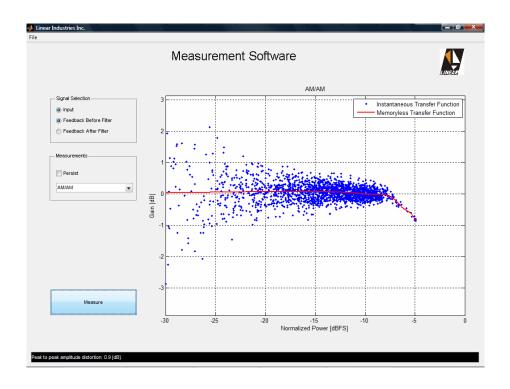
Amplitude distortion via AM/AM, AM/PM and CCDF







AM/AM

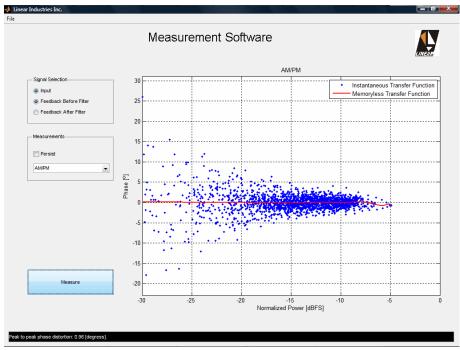


AM/AM: Amplitude Modulation / Amplitude Modulation.

- Should be measured after amplifier and before filter
- Represents normalized output/input amplitude ratio (amplifier gain) in [dB] vs. normalized input amplitude in [dB] full scale
- This evaluates a power amplifiers transfer function and linearity characteristics



AM/PM

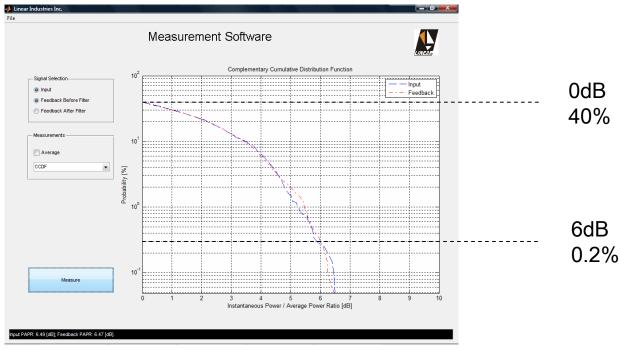


AM/PM: Amplitude Modulation / Phase Modulation

- Should be measured after amplifier and before filter
- Represents the phase deviation in [°] vs. normalized input amplitude in [dB]
- This evaluates a power amplifiers transfer function and its phase characteristics



CCDF



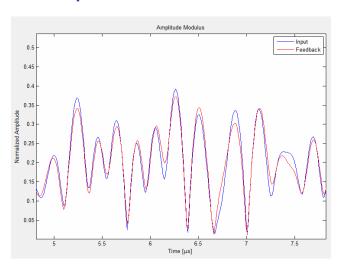
CCDF: Complementary Cumulative Distribution Function (CCDF)

- Represents the probability of instantaneous power to be above average power. If amplifier in compression, red line would be below blue line.
- Use this measurement to evaluate amplifier compressed or clipping PAPR (Peak to Average Power Ratio) measurement represents the ratio between the maximum peak power and average power
- Example: 0dB = 40 % of the time > 10kW or 60 % of the time < 10kW
 +6dB = 0.2 % of the time > 40kW or 99.8% of the time < 40kW

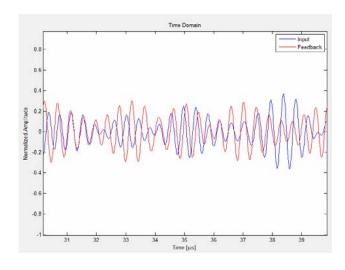


I and Q errors via

Amplitude Modulus



Time Domain





I and Q errors

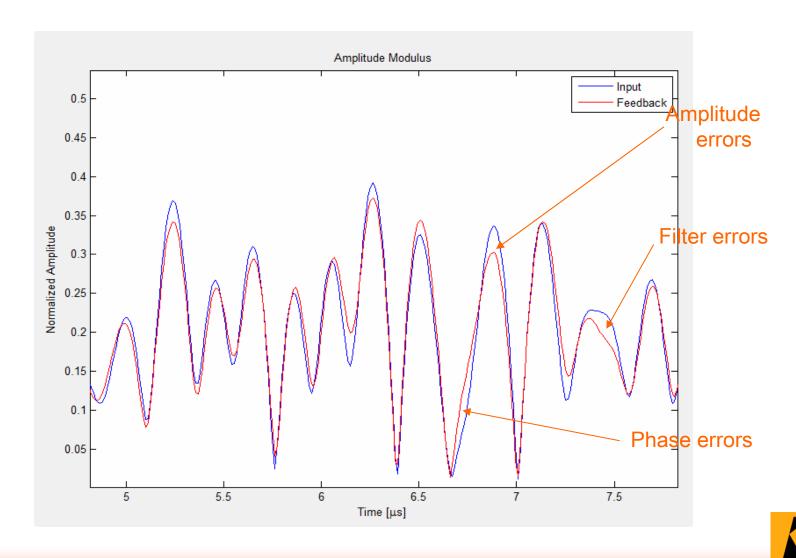
- I = real part (In-phase component)
- Q = imaginary part (quadrature component).
- Each sample as a vector on a plane, the amplitude modulus is the magnitude of this vector:

$$MOD = \sqrt{(I^2 + Q^2)}$$

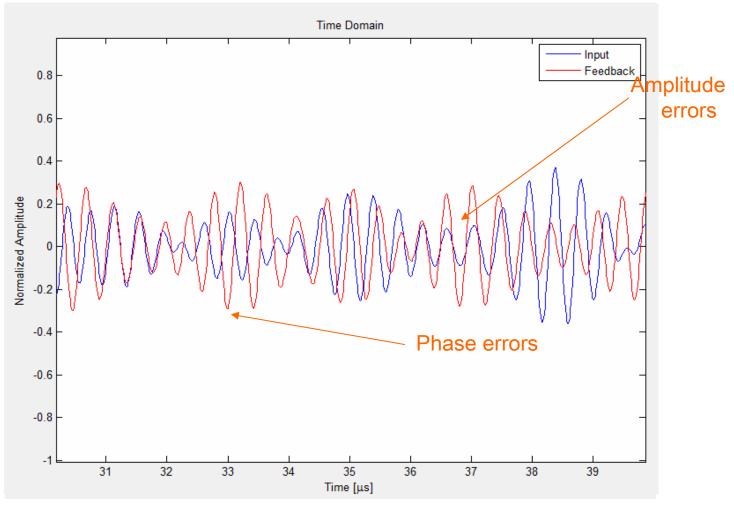
- Two ways to represent a complex signal and the I Q errors:
 - Real part (I) of signal (Cartesian coordinates), wrt the imaginary part
 - Real and Imaginary (I and Q)
 - Polar coordinates Amplitude (Modulus) only
 - Modulus and Angle (I only)



I and Q Errors (Amplitude Modulus)

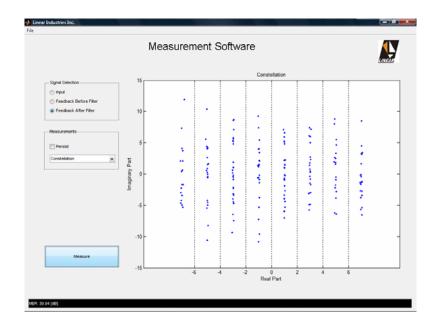


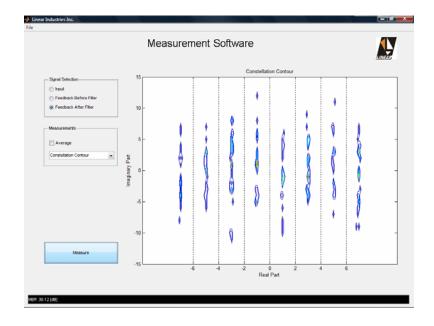
I and Q Errors (Time Domain)





ATSC Constellation and Constellation Contour

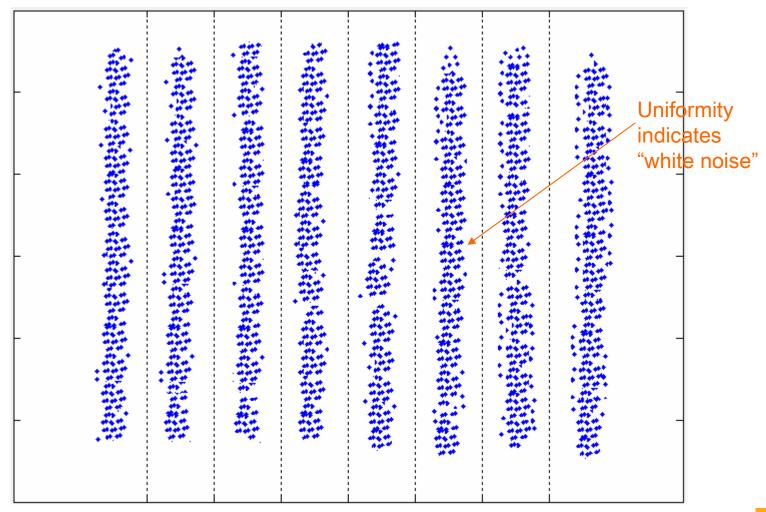




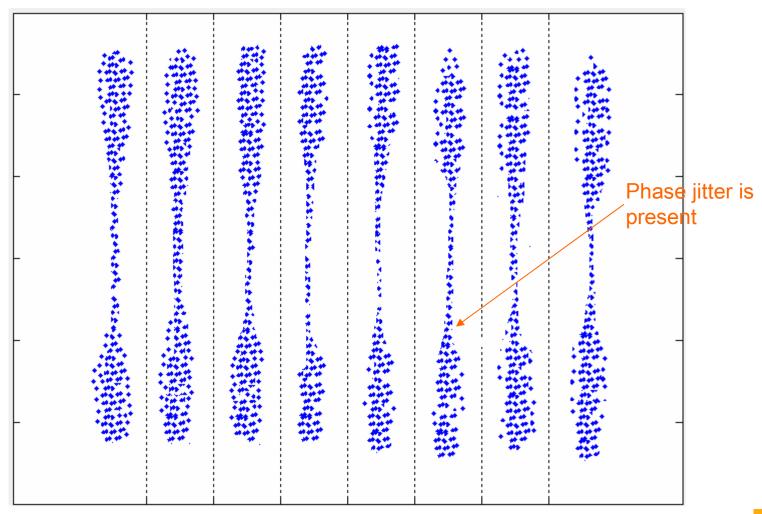


- A two dimensional graphical representation of the 8-VSB RF carrier ampl. and phase at each sampling time
- The 8 VSB constellation diagram shows 8 lines (whereas 16 or 64 QAM would show 16 or 64 points)
- The 8VSB constellation diagram evaluates noise impairments such as SNR/MER.
- The thinner the lines the lower the distortion
- With "pure" noise impairment the lines are widened over their entire length
- The RMS value of the noise can be determined by constellation analysis and using a Gaussian (normal) distribution statistical function
- This determines the IQ "decision points" standard deviation from the ideal point

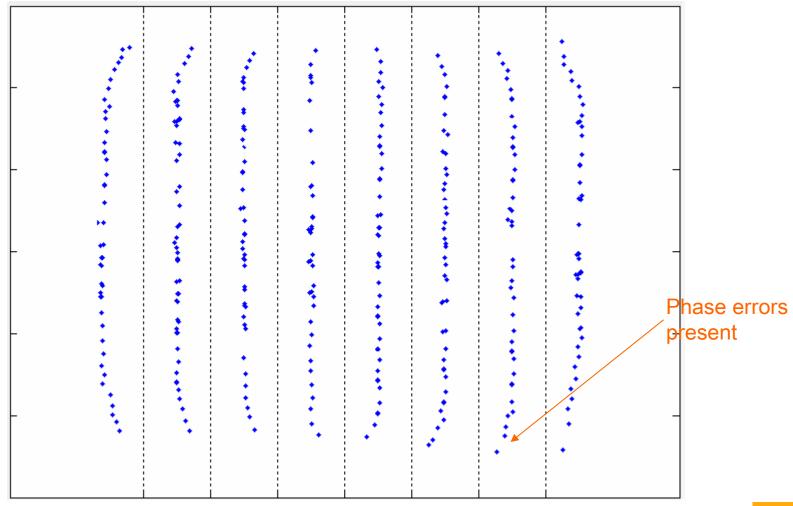






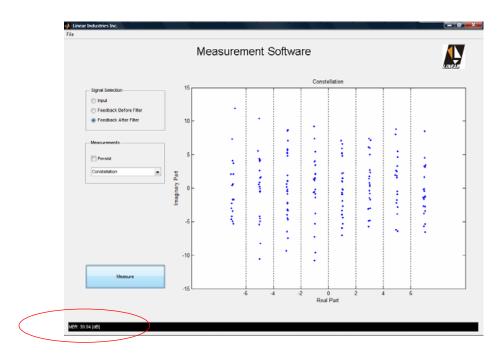








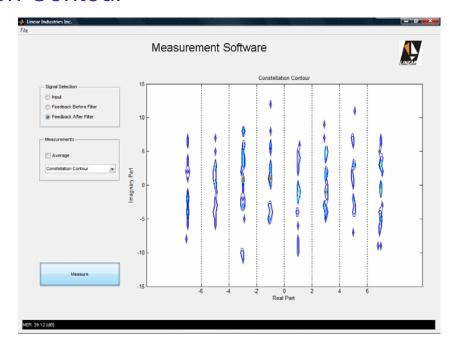
ATSC Constellation



Constellation

- A representation of the sampled baseband modulating symbols, where the Cartesian plan is
 plotted by its real vs. imaginary part. The real part (in-phase) of the sampled symbols carry
 the transmitted information while the imaginary part (quadrature) carries the necessary
 information to generate the vestigial side band. For 8VSB modulation the eight regions
 represent the eight possible symbols and the seven dashed lines indicate the decision
 threshold.
- The Modulation Error Ratio (MER) is provided in [dB]
- ► lin·e·ar direct, to the point; advanced technology, cost-efficient economics.

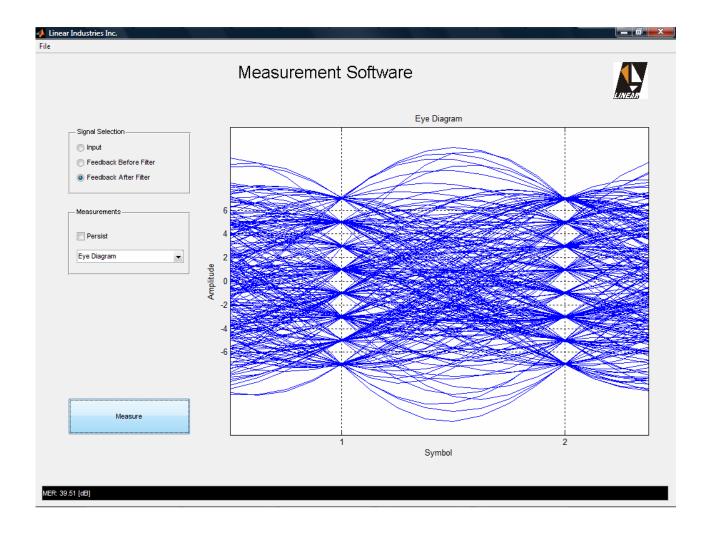
ATSC Constellation Contour



Constellation Contour

- Shows probability regions for a demodulated symbol to fail in coded in colors. A red area means low probability of failure, the blue area means higher probability of failure
- This measurement reflects the signal quality. Use it to measure non-linear, linear and amplitude distortions
- MER is the mean squared error in [dB] assuming the perfect symbol as the reference signal







- An overlay of many received RF amplitude signal traces at the instant of sampling – the RF signal must attain one of 8 levels
- Can be used to evaluate many types of errors
- The convergence of "perfect" signal traces forms seven "eyes" that coincide with the occurrence of the receiver clock pulses
- Errors can be caused by noise, low-image rejection ratio, phase noise, carrier suppression and non linear and linear distortion

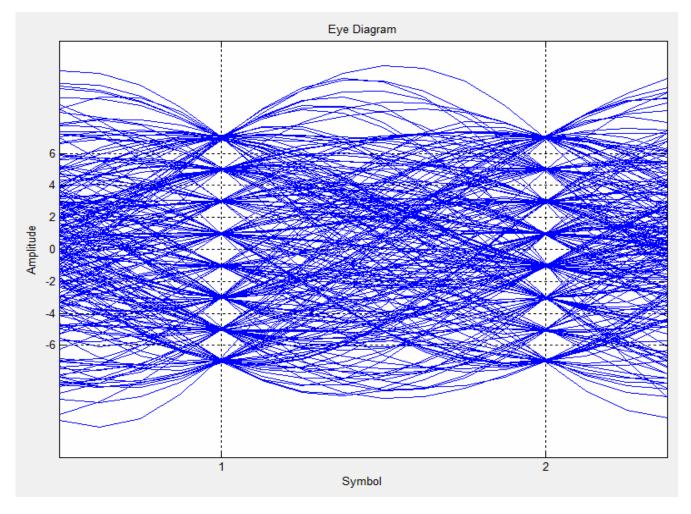


- For each type of error, an error vector is continually calculated, and the sum of the squares (RMS value) of all error vectors is calculated
- The ratio of the error-vector RMS value to the signal amplitude produces the value is the MER [specified in dB]
- The quality of the in-channel emitted signal can be specified and measured by determining the departure from 100 percent "eye" opening
- The departure, or error, has four identifiable components: 1) circuit or "white" noise, 2) inter-modulation noise caused by non-linearity,
 3) inter-symbol interference, and 4) phase noise

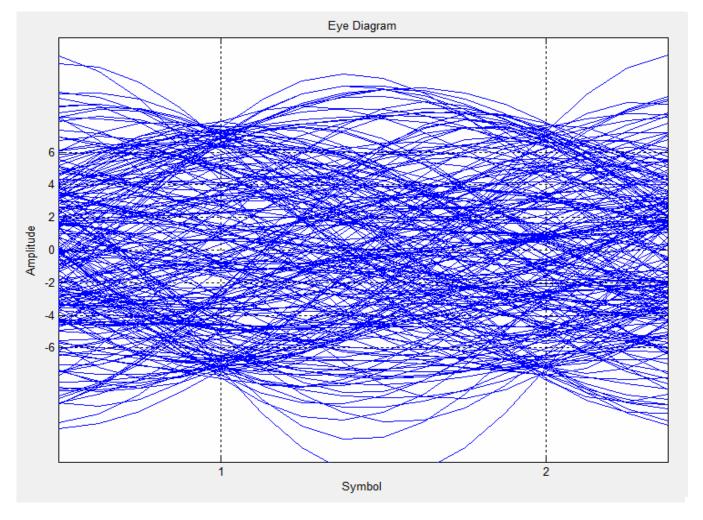


- The combination of all of these effects can be specified and measured by an error (or noise) magnitude power. The error magnitude power for a DTV signal should be no greater than – 27 dB relative to the authorized signal power.
- An error power of this level will increase the received DTV error threshold by a maximum value of about 0.25 dB, which for UHF assignments corresponds to a reduction of approximately 1/4 mile in coverage distance from the transmitter
- The error is the total power of all the above errors along the inphase demodulation axis, and is considered as "noise" when measuring signal-to-noise ratio



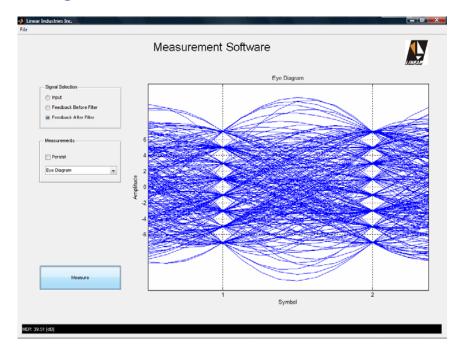








MER and the "Eye" diagram



Eye Diagram:

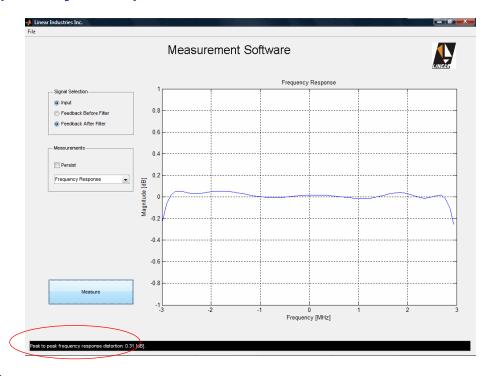
- Represents the pure "real" modulating symbols filtered with a raised cosine filter and persisted
 with a symbol period multiple. The open "eyes" indicate the exact moment for the receiver to
 sample and decide the level of the symbol in order to extract its value
- Use it to realize non-linear, linear and amplitude distortions. It can be measured before and after the filter to understand the individual and cumulative distortions of both parts of the system

Amplitude Response and Group Delay

- The 8VSB signal has no pilot signals that provide information about channel quality however...
- Amplitude response and group delay can still used to align a modulator or transmitter or determine where a fault may be present
- A in band frequency response and Group delay spectrum are valuable tools to evaluate the characteristics of an 8VSB signal
 - Shoulder attenuation
 - Amplitude frequency response
 - Pilot carrier amplitude



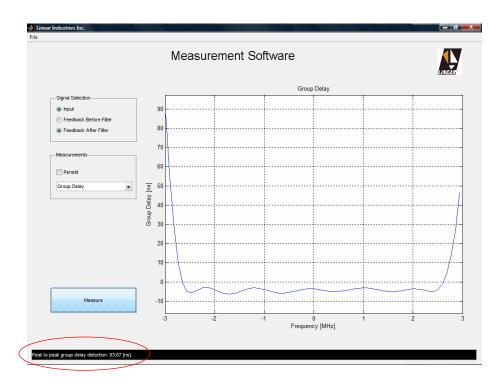
Amplitude Frequency Response



Frequency Response

- This measurement will show the in-band frequency response of the device under test (DUT).
 This measurement can be used to measure amplitude versus frequency distortions of such devices as amplifiers, combiners, couplers, filters and feedback systems.
- Peak to peak frequency response distortion: difference in amplitude in [dB] (p-p). Ideal response should be less than 1.5[dB].

Group Delay



Group Delay

- This measurement will show the in-band Group delay response of the device under test (DUT). This measurement can be used for linear distortion of devices such as combiners, feedback circuits and filtering networks. Typically most linear distortion is caused by the output mask filter and therefore the measurement should be taken after filter
- Peak to peak group delay: difference in the delay in [dB] (p-p). Ideal delay should be less than 150[nS].



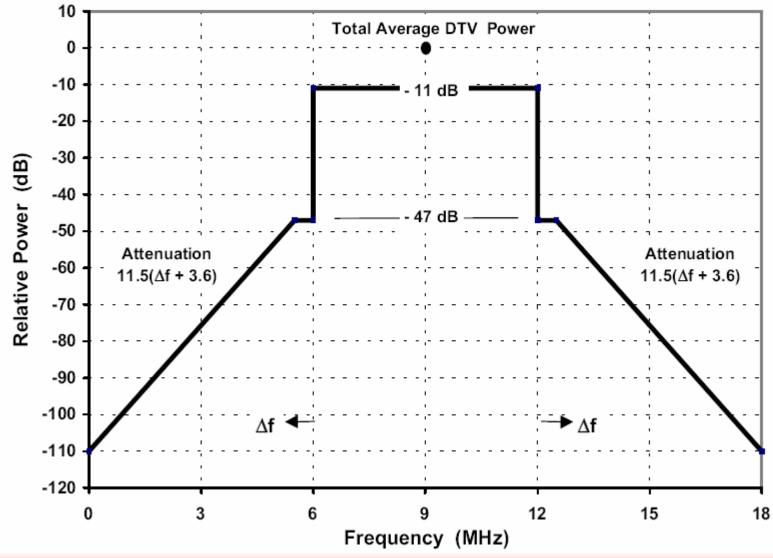
Shoulder Measurement via the Power Spectrum

- The DTV emission mask was adopted in FCC Memo O/R requires that:
- In the first <u>500 kHz</u> from the authorized channel edge, transmitter emissions must be attenuated no less than <u>47 dB</u> below the average transmitted power;
- More than 6 MHz from the channel edge, emissions must be attenuated no less than 110 dB below the average transmitted power;
- At any frequency between 0.5 and 6 MHz from the channel edge, emissions must be attenuated no less than the value determined by the following formula: Attenuation in dB = 11.5 (Δ f + 3.6) where Δ f = frequency difference in MHz from the edge of the channel

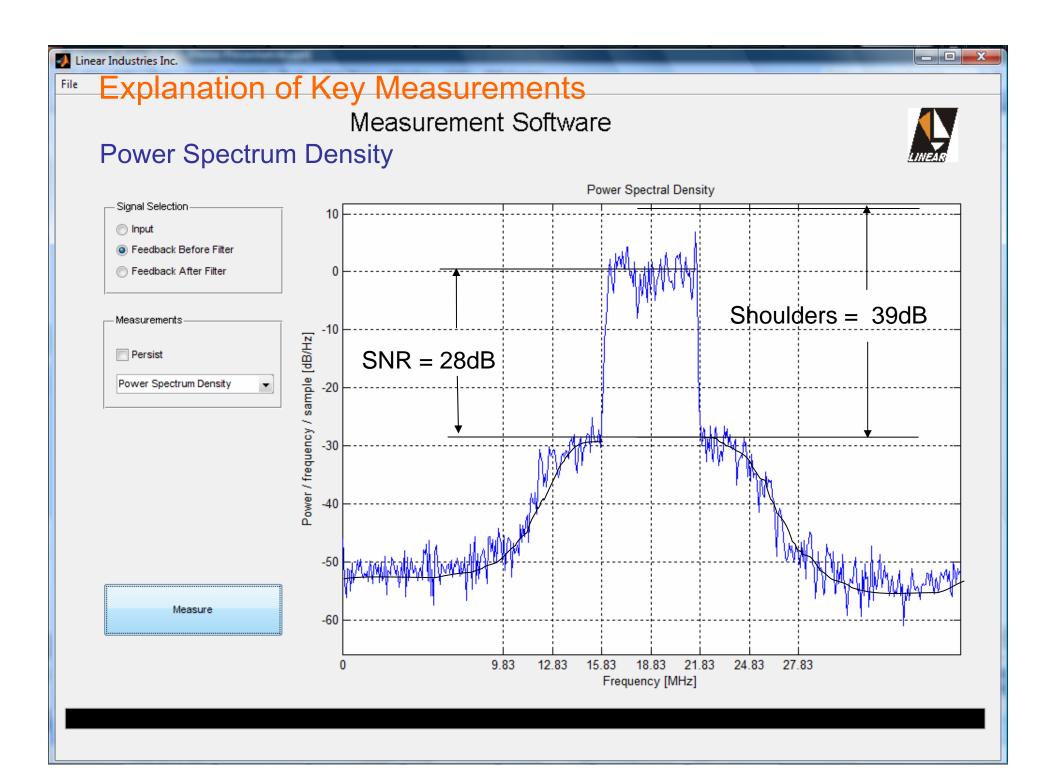


Explanation of Key Measurements

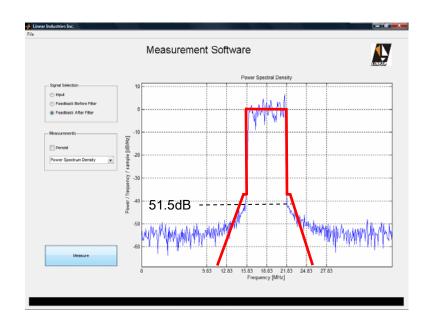
Power Spectrum Density

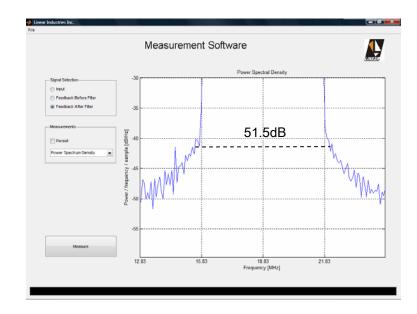






Power Spectrum Density measurement





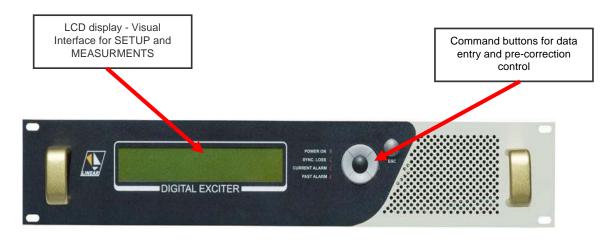
Power Spectrum Density

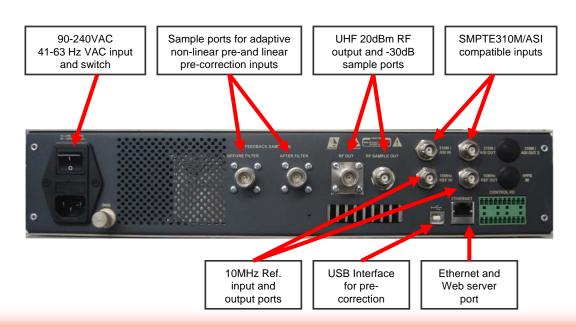
This is a spectral analysis using the Power Spectrum Density estimation of the output signal. The in-band signal is normalized to be 0 [dB] point. The shoulders can be measured at +/-3MHz from the authorized channel edges. The Average power measured in the first 500kHz must be less than -47dB with respect to -11dB at 500kHz resolution.



2. Correction demonstration







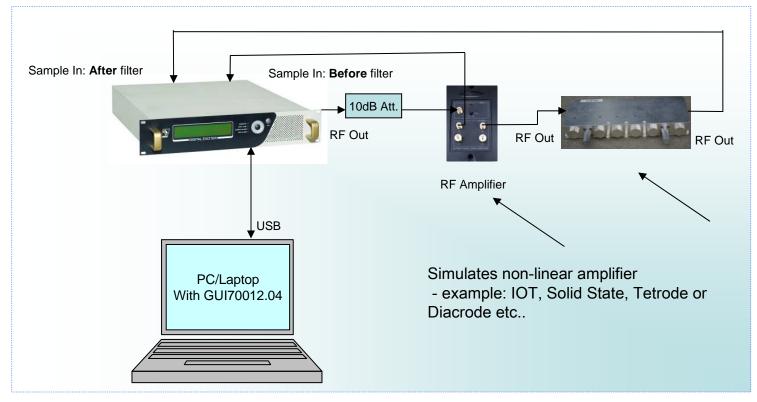


Conditions

Exciter Channel: 44 RF Output: +1dBm Amplifier: RS Cable

ATSC Mask Filter channel 44

Laptop PC with Windows Vista and GUI7xxx.xx software





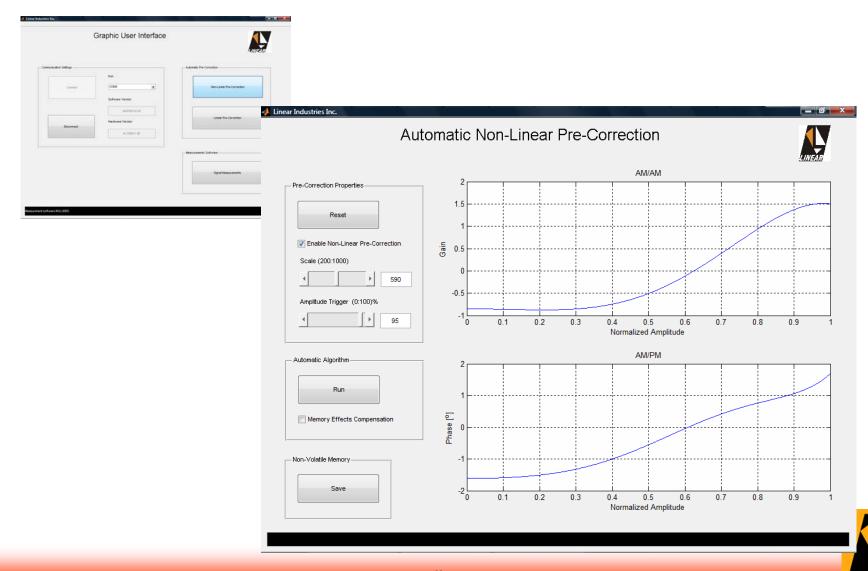
Amplifier Performance @ min. and Max. gain

Measurement		@ min. Gain	@ max. Gain	Diff.
Compression	dB	0.54	1.96	-1.42
Amplitude Distortion	dB	0.54	3.07	-2.53
Phase Distortion	Degrees	1.62	3.79	-2.17
PAPR Difference	dB	0.21	1.92	-1.71
MER	dB	33	26	-7
Frequency Response	dB	0.93	2.03	-1.1
Group Delay	nS	121	224*	-103
Shoulder Levels	dB	47	39	-8

^{*} After MASK filter

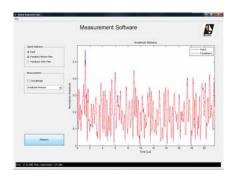


Run non-linear correction ("at the push of a button")

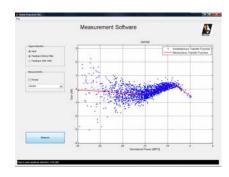


Performance before filter <u>after</u> non-linear correction

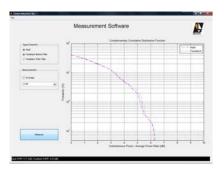
0.68dB compression



0.83dB amplitude distortion



6.31/6.25dB PAPR



35.86dB MER



35.86dB MER

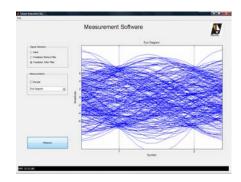


48dB Shoulders

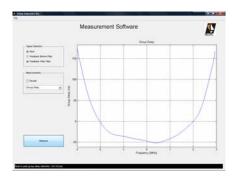


Performance <u>after</u> filter <u>without</u> linear correction

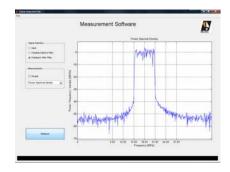
20.35 MER



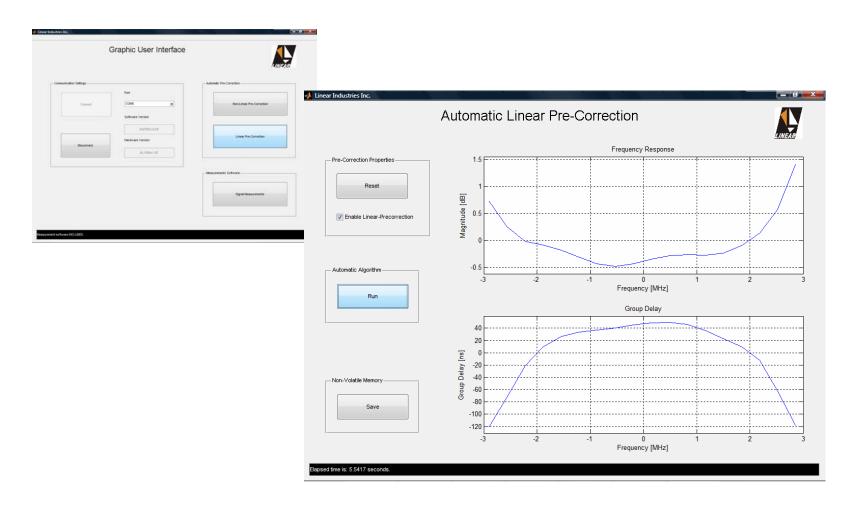
224nS Group Delay



48dB Shoulders



"Run" linear correction ("at the push of a button")





Final Performance results

Measurement	_	ax. Gain Before ection	@ max Correct	. Gain After tion	Improvement.
Compression	dB	1.96	C).32	1.64
Amplitude Distortion	dB	3.07	C).4	2.53
Phase Distortion	Deg.	3.79	C).51	3.28
PAPR Difference	dB	1.92	C).24	1.68
MER	dB	26	3	38	12
Frequency Response	dB	2.03	C).6	1.43
Group Delay	nS	224*	7	79	145
Shoulder Levels	dB	39	5	51	12

^{*} After MASK filter

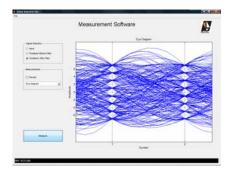


Delineation - clarity, distinctness, exactness

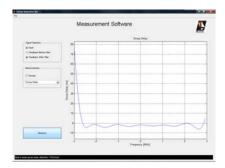


Final Performance results

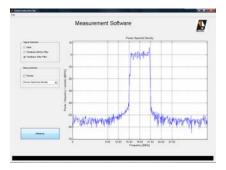
38.03dB MER



79nS Group Delay



51dB Shoulders



Summary

- With automatic correction almost any OEM transmitter and any OEM filter can be corrected to meet the FCC recommended performance
- Shoulders of 50+ dB and MER of 36+ dB obtainable (well within spec!)
- Not necessary to spend \$50,000+ on an ATSC automatic correcting exciter
- One product corrects, monitors and measures the performance of any the transmitter for less than \$25k*
- Low power solid state transmitters can be upgraded to digital for less than \$10k*



^{*}price exclusive of taxes, shipping and handling. Additional options available

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